

SOLUTION OF EXAM 01
Subject: CONTROL THEORY

Exercise 1 (5 marks):

a) (1.5 marks)

i. (0.5 marks) Consider the Polynomial:

$$A(s) = T_2^2 s^3 + 2T_2 s^2 + s + k_1 k = 4s^3 + 4s^2 + s + 0.5k_1$$

ii. (0.5d) Routh Table

4	1
4	$0.5k_1$
$\frac{4 - 2k_1}{4}$	
$0.5k_1$	

We obtain $0 < k_1 < 2$

iii. (0.5 d)

Because of stable closed loop system, there exists the Limitation

$$\lim_{t \rightarrow \infty} (u(t) - y(t)) = \lim_{s \rightarrow 0} s(U(s) - Y(s)) = \lim_{s \rightarrow 0} sU(s) \left(1 - \frac{k_2 k_1 G(s)}{1 + k_1 G(s)} \right);$$

$$u(t) = 1(t) \text{ v\`a } \lim_{s \rightarrow 0} G(s) = +\infty, \text{ we obtain } \lim_{t \rightarrow \infty} (u(t) - y(t)) = 1 - k_2$$

We imply $k_2 = 1$;

b) (2 marks)

i. (1 mark) $G(s) = \frac{k}{s(1 + T_2 s)^2}$

$$a = 4: T_I = T_1 + 4T_2, k_p = \frac{T_I}{8kT_2^2}, T_D = \frac{4T_1 T_2}{T_I}, T = 4T_2 \text{ with}$$

$$k = 0.5, T_1 = T_2 = 2 \text{ we obtain } T_I = 10, k_p = \frac{5}{8}, T_D = 1.6, T = 8$$

ii. (1 mark) Stability reserve of closed system does not depend on $R_2(s)$. We need to obtain the open system with transfer function:

$$\begin{aligned} G_h(s) &= R_1(s)G(s) = k_p \left(1 + \frac{1}{T_I s} + T_D s \right) \cdot \frac{k}{s(1 + T_1 s)(1 + T_2 s)} \\ &= \frac{k_p(1 + T_A s)(1 + T_B s)}{T_I s} \cdot \frac{k}{s(1 + T_1 s)(1 + T_2 s)} = \frac{k_p k(1 + T_B s)}{T_I s^2(1 + T_2 s)} \end{aligned}$$

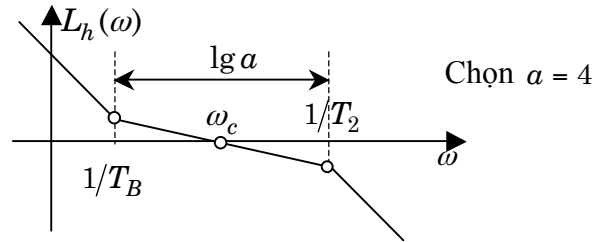
We select $T_A = T_1$ with $T_A + T_B = T_I$, $T_A T_B = T_I T_D$, $T_B = 4T_2 > T_2$, we obtain the Bode diagram of open system as follows.

$$\omega_c = \frac{1}{\sqrt{T_B T_2}}$$

$T_B = 8, T_1 = T_2 = 2$ we imply $\omega_c = \frac{1}{4}$. Stability reserve of closed system $\Delta\varphi$:

$$\Delta\varphi = -\pi - \varphi_c = -\pi - \arctan G_h(j\omega_c) = \arctan(\omega_c T_2) - \arctan(\omega_c T_B)$$

$$\Delta\varphi = \arctan\left(\frac{1}{2}\right) - \arctan(2)$$



c) (1.5 marks)

- i. (0.5 marks) Using the Theorem “The input is sinusoidal signal, the output comes to sinusoidal signal depending on $G_k(j\omega)$ ”;
- ii. (0.5 marks) If $R_1(s)$ is PI Controller then we do not obtain $G_k(j\omega) = 1$;
- iii. (0.5 marks) $R_1(s) = \frac{a}{s^2 + \omega^2} (a > 0)$

Exercise 2

a) (1 marks) we denote

$$A = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 2 & 1 \\ 0 & 1 & 1 \end{pmatrix}, B = \begin{pmatrix} 1 \\ 0 \\ 1 \end{pmatrix}, \underline{c} = \begin{pmatrix} a \\ 0 \\ 1 \end{pmatrix} \Rightarrow \begin{cases} \frac{dx}{dt} = Ax + Bu \\ y = \underline{c}^T x \end{cases}$$

i. (0.5 marks)

The equivalent polynomial of matrix A will be:

$$\det(sI - A) = (s - 1)(s^2 - 3s + 1). \text{ This system is not stable;}$$

ii. (0.5 marks) $\text{Rank}(B, AB, A^2B) = 3 \Rightarrow$ This system is controllable

b) (1 marks)

i. (0.5 marks)
$$N = \begin{pmatrix} \underline{c}^T \\ \underline{c}^T A \\ \underline{c}^T A^2 \end{pmatrix} = \begin{pmatrix} a & 0 & 1 \\ a & 1 & a+1 \\ a & a+3 & 2a+2 \end{pmatrix}$$

ii. (0.5 marks) $\det(N) = -a(a^2 + a - 1)$

In order to obtain the observability property: $\det(N) \neq 0 \Leftrightarrow a \neq 0; \frac{-1 \pm \sqrt{5}}{2}$

c) (2 marks)

i. (1 marks) State Feedback control is $\underline{u} = \underline{w} - R\underline{x}$ với $R = (r_1 \ r_2 \ r_3)$ satisfy eigen values of $(A - \underline{b}r^T)$ inside $(-2, 0)$ and we select all of eigen values

being -1), we obtain based on Ackermann

$$[r_1 \ r_2 \ r_3] = [0 \ 0 \ 1] M^{-1} \Phi_R(A) = [-4, -24, 11]$$

ii. (1 marks) Observer

i. (0.5 marks) $\underline{\dot{x}}$ is the root of difference equation

$$\frac{d\underline{x}}{dt} = A\underline{x} + \underline{b}u + L(y - \underline{c}^T \underline{x}).$$

ii. (0.5 marks) Find matrix L to obtain that $(A - L\underline{c}^T)$ have all eigenvalues being -3 (faster than e^{-2t}). Using Ackermann:

$$L = \Phi_L(A)N^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} = [28, 77, -15]^T$$

(1d) Drawing. The closed system is not controllable because $\underline{\dot{x}}$ converge to \underline{x}

SOLUTION OF EXAM 02
Subject: CONTROL THEORY

Exercise 1 (5 marks):

d) (1.5 marks)

d) (0.5 marks) Consider the Polynomial:

$$A(s) = T_2 s^3 + 2T_2 s^2 + s + k_1 k = s^3 + 2s^2 + s + 10k_1$$

i. **(0.5d)** Routh Table

1	1
2	$10k_1$
$\frac{2 - 10k_1}{2}$	
$10k_1$	

We obtain: $0 < k_1 < 0.2$

ii. **(0.5 d)**

Because of stable closed loop system, there exists the Limitation

$$\lim_{t \rightarrow \infty} (u(t) - y(t)) = \lim_{s \rightarrow 0} s(U(s) - Y(s)) = \lim_{s \rightarrow 0} sU(s) \left(1 - \frac{k_2 k_1 G(s)}{1 + k_1 G(s)} \right);$$

Because $u(t) = 1(t)$, we imply $\lim_{s \rightarrow 0} G(s) = +\infty$ and

$$\lim_{t \rightarrow \infty} (u(t) - y(t)) = 1 - k_2, \text{ we obtain } k_2 = 1;$$

b) (2 marks)

i. **(1 marks)** $G(s) = \frac{k}{s(1 + T_2 s)^2}$

$(a = 4): T_I = T_1 + 4T_2, k_p = \frac{T_I}{8kT_2^2}, T_D = \frac{4T_1 T_2}{T_I}, T = 4T_2$ where

$k = 10, T_1 = T_2 = 1$ we obtain $T_I = 5, k_p = \frac{1}{16}, T_D = 0.8, T = 4$

i. **(1 mark)** Stability reserve of closed system does not depend on $R_2(s)$. We

need to obtain the open system with transfer function:

$$\begin{aligned} G_h(s) &= R_1(s)G(s) = k_p \left(1 + \frac{1}{T_I s} + T_D s \right) \cdot \frac{k}{s(1 + T_1 s)(1 + T_2 s)} \\ &= \frac{k_p(1 + T_A s)(1 + T_B s)}{T_I s} \cdot \frac{k}{s(1 + T_1 s)(1 + T_2 s)} = \frac{k_p k(1 + T_B s)}{T_I s^2(1 + T_2 s)} \end{aligned}$$

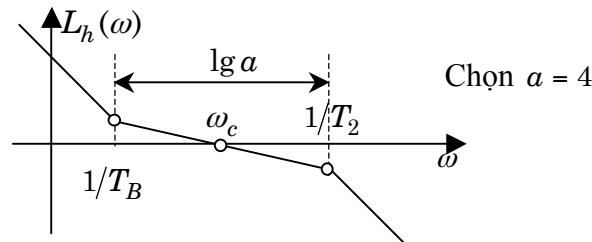
We select $T_A = T_1$ with $T_A + T_B = T_I, T_A T_B = T_I T_D, T_B = 4T_2 > T_2$, we obtain the Bode diagram of open system as follows.

$$\omega_c = \frac{1}{\sqrt{T_B T_2}}.$$

$T_B = 8, T_1 = T_2 = 2$ we imply $\omega_c = \frac{1}{4}$. Stability reserve of closed system $\Delta\varphi$:

$$\Delta\varphi = -\pi - \varphi_c = -\pi - \arctan G_h(j\omega_c) = \arctan(\omega_c T_2) - \arctan(\omega_c T_B)$$

$$\Delta\varphi = \arctan\left(\frac{1}{2}\right) - \arctan(2)$$



(1.5 marks)

- a. (0.5 marks) Using the Theorem “The input is sinusoidal signal, the output comes to sinusoidal signal depending on $G_k(j\omega)$ ”;
- b. (0.5 marks) If $R_1(s)$ is PI Controller then we do not obtain $G_k(j\omega) = 1$;
- ii. (0.5 marks) $R_1(s) = \frac{a}{s^2 + \omega^2} (a > 0)$

Exercise 2

a) (1 marks) we denote

$$A = \begin{pmatrix} 1 & 0 & 1 \\ 0 & 2 & 1 \\ 0 & 1 & 1 \end{pmatrix}, B = \begin{pmatrix} 1 \\ 0 \\ 1 \end{pmatrix}, \underline{c} = \begin{pmatrix} a \\ 0 \\ 1 \end{pmatrix} \Rightarrow \begin{cases} \frac{dx}{dt} = Ax + Bu \\ y = \underline{c}^T x \end{cases}$$

i. (0.5 marks)

The equivalent polynomial of matrix A will be:

$$\det(sI - A) = (s - 1)(s^2 - 3s + 1). \text{ This system is not stable;}$$

ii. (0.5 marks) $\text{Rank}(B, AB, A^2B) = 3 \Rightarrow$ This system is controllable

b) (1 marks)

$$i. \quad (0.5 \text{ marks}) \quad N = \begin{pmatrix} \underline{c}^T \\ \underline{c}^T A \\ \underline{c}^T A^2 \end{pmatrix} = \begin{pmatrix} a & 0 & 1 \\ a & 1 & a+1 \\ a & a+3 & 2a+2 \end{pmatrix}$$

$$ii. \quad (0.5 \text{ marks}) \quad \det(N) = -a(a^2 + a - 1)$$

In order to obtain the observability property: $\det(N) \neq 0 \Leftrightarrow a \neq 0; \frac{-1 \pm \sqrt{5}}{2}$

c) (2 marks)

i. (1 marks) State Feedback control is $\underline{u} = \underline{\omega} - R\underline{x}$ với $R = (r_1 \ r_2 \ r_3)$ satisfy eigen values of $(A - \underline{b}r^T)$ inside $(-2, 0)$ and we select all of eigen values

being -1), we obtain based on Ackermann

$$[r_1 \ r_2 \ r_3] = [0 \ 0 \ 1] M^{-1} \Phi_R(A) = [-4, -24, 11]$$

ii. (1 marks) Observer

i. (0.5 marks) $\underline{\hat{x}}$ is the root of difference equation

$$\frac{d\underline{\hat{x}}}{dt} = A\underline{\hat{x}} + \underline{b}u + L(y - \underline{c}^T \underline{\hat{x}}).$$

- ii. **(0.5 marks)** Find matrix L to obtain that $(A - L\underline{c}^T)$ have all eigenvalues being -3 (faster than e^{-2t}). Using Ackermann:

$$L = \Phi_L(A)N^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} = [28, 77, -15]^T$$

(1d) Drawing. The closed system is not controllable because \underline{x} converge to \underline{x}